

Ruofeng YU, Chenyang LUO, Mengdi BAI, Shangqu YAN, Wei YANG, Yaowen FU, 2025. Waveform design based on mutual information upper bound for joint detection and estimation. *Frontiers of Information Technology & Electronic Engineering*, 26(11):2324-2337. <https://doi.org/10.1631/FITEE.2500276>

Waveform design based on mutual information upper bound for joint detection and estimation

Key words: Radar waveform design; Mutual information upper bound; Target detection; Parameter estimation; Constant modulus constraint

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Motivation

1. In practice, detection and estimation are interdependent, motivating joint optimization.
2. Multi-objective formulations provide partial solutions, while learning-based approaches suffer from complexity and generalization issues.
3. Information-theoretic designs have been explored, e.g., a two-stage scheme using mutual information (MI) and Kullback–Leibler (KL) divergence, yet a unified foundation remains lacking.

Main idea

1. Unified variational framework: Joint detection-estimation waveform design is formulated via mutual information upper bound (MIUB) maximization, unifying MI- and KL-based criteria.
2. Flexible statistical modeling: Gaussian mixture distributions (GMDs) are adopted for target and clutter, combining fidelity with analytical tractability.
3. Phase-coded beam optimization algorithm (PC-DOA): A constant modulus synthesis algorithm with hybrid initialization and adaptive dynamics is developed to enhance convergence and solution quality.

Method

1. The received signal y is modeled by the following binary hypothesis test.

$$\begin{cases} \mathcal{H}_0: & y = w, \\ \mathcal{H}_1: & y = Sx + w, \end{cases}$$

The corresponding likelihoods of y under two hypothesis:

$$p_0(y) \triangleq p(y|\mathcal{H}_0) = \sum_{k=1}^K \alpha_k \mathcal{CN}(y; 0, R_k),$$

$$p_1(y) \triangleq p(y|\mathcal{H}_1) = \sum_{\ell=1}^L \gamma_\ell \mathcal{CN}(y; 0, \Sigma_\ell),$$

The constant modulus constraint:

$$|s_n| = c = \sqrt{E_s / N}, \quad n = 1, 2, \dots, N,$$

MIUB decomposition:

$$\text{MIUB}(s) = \underbrace{I(x; y)}_{E(s)} + \underbrace{\mathcal{D}_{\text{KL}}(p_1(y) \square p_0(y))}_{D(s)},$$

Method (Cont'd)

2. The exact evaluation of MIUB under GMD likelihoods is generally intractable. Tractable approximation strategies: to circumvent this computational challenge, we adopt the following tractable approximation strategies.

➤ MI approximation (Lemma 2): the MI term can be approximated as $\bar{E}(s) = \Xi(\{\alpha_k, R_k\}_{k=1}^K) - \Xi(\{\gamma_\ell, \Sigma_\ell\}_{\ell=1}^L)$,

➤ KL divergence approximation (Lemma 3): the KL divergence term admits the approximation $\bar{D}(s) = \sum_{\ell=1}^L \gamma_\ell \left[\ln \frac{\gamma_\ell}{\alpha_{k^*(\ell)}} + \mathcal{L}(\Sigma_\ell, R_{k^*(\ell)}) \right]$,

Final optimization problem (P1):

$$(\mathcal{P}_1): \begin{array}{ll} \max_s & F(s) \square \bar{E}(s) + \bar{D}(s) \\ \text{s.t.} & |s_n| = c, \quad n = 1, 2, \dots, N \end{array}$$

Method (Cont'd)

3. We propose the PC-DOA, a metaheuristic specifically designed for MIUB-based waveform design under constant modulus constraints.

Phase-coded representation: $s(\theta) = c \cdot [e^{j\theta_1}, e^{j\theta_2}, \dots, e^{j\theta_N}]^T$.

Hybrid initialization strategy:

➤ LFM-inspired initialization: $[\theta_{\text{LFM},i}^0]_n = \Psi\left(\beta_i \pi \frac{(n-1)^2}{N-1} + \Delta_{i,n}\right)$,

➤ Chaotic initialization: $\rho_n^k = 4\rho_n^{k-1}(1-\rho_n^{k-1})$, $[\theta_{\text{chaos},i}^0]_n = \Psi\left(-\pi + 2\pi\rho_n^{K_{\text{chaos}}}\right)$.

➤ Random initialization: $[\theta_{\text{rand},i}^0]_n \sim \mathcal{U}(-\pi, \pi)$.

Exploration and exploitation:

➤ Exploration phase: $\theta_g^{\text{best}} = \arg \max_{\theta \in \mathcal{G}_g} f(\theta)$.

➤ Exploitation phase: $[\theta_{\text{new}}]_n = \Psi\left([\theta^{\text{global}}]_n + \eta(t) \cdot \Delta_n\right)$, $n \in \mathcal{I}$.

Method (Cont'd)

4. The PC-DOA algorithm initializes a population using a hybrid strategy and finds the initial best solution θ^* . It then iterates T times, splitting the search into two phases: first, a broad exploration phase, followed by an exploitation phase that focuses the search around the current θ^* . In each iteration, it evaluates the new candidates, selects the next generation, and updates θ^* if a better solution is found. Finally, it computes and returns the optimal waveform s^* based on the final θ^* .

Algorithm 1 Phase-coded dream optimization algorithm

Input: Objective function $F : \mathcal{M} \rightarrow \mathbb{R}$, wave-form parameters $\{N, c\}$, and algorithm parameters $\{N_p, T, G, \eta, \alpha\}$

Output: Optimal phase vector θ^* and waveform s^*

```
1: Initialize Population  $\mathcal{P}_0 = \{\theta_1^0, \theta_2^0, \dots, \theta_{N_p}^0\}$  via hybrid strategy Eqs. (21)–(24), fitness  $f(\theta) \leftarrow F(\mathbf{s}(\theta))$  for all  $\theta \in \mathcal{P}_0$ , global best  $\theta^* \leftarrow \arg \max_{\theta \in \mathcal{P}_0} f(\theta)$ , and  $f^* \leftarrow f(\theta^*)$ 
2: for  $t = 1$  to  $T$  do
3:   Initialize  $\mathcal{P}_{\text{candidate}} \leftarrow \emptyset$ 
   //  $\mathcal{P}_{\text{candidate}}$  denotes the candidate population
4:   if  $t \leq \alpha T$  then
5:     Partition  $\mathcal{P}_{t-1}$  into  $G$  subgroups
6:     for each  $\theta_{\text{old}} \in \mathcal{P}_{t-1}$  do
7:       Generate  $\theta_{\text{new}}$  from  $\theta_{\text{old}}$  via Eq. (28)
8:        $\mathcal{P}_{\text{candidate}} \leftarrow \mathcal{P}_{\text{candidate}} \cup \{\theta_{\text{new}}\}$ 
9:     end for
10:  else
11:    for each  $\theta_{\text{old}} \in \mathcal{P}_{t-1}$  do
12:      Generate  $\theta_{\text{new}}$  guided by  $\theta^*$  via Eq. (30)
13:       $\mathcal{P}_{\text{candidate}} \leftarrow \mathcal{P}_{\text{candidate}} \cup \{\theta_{\text{new}}\}$ 
14:    end for
15:  end if
16:  Evaluate fitness  $f(\theta)$  for all  $\theta \in \mathcal{P}_{\text{candidate}}$ 
17:  Select next-generation population  $\mathcal{P}_t$  from  $\mathcal{P}_{t-1} \cup \mathcal{P}_{\text{candidate}}$ 
18:  if  $\max_{\theta \in \mathcal{P}_t} f(\theta) > f^*$  then
19:     $\theta^* \leftarrow \arg \max_{\theta \in \mathcal{P}_t} f(\theta)$ 
20:     $f^* \leftarrow f(\theta^*)$ 
21:  end if
22: end for
23: Compute optimal waveform  $s^* \leftarrow c \cdot \exp(j\theta^*)$ 
24: Return  $\theta^*$  and  $s^*$ 
```

Major results

Comparison of optimization objectives

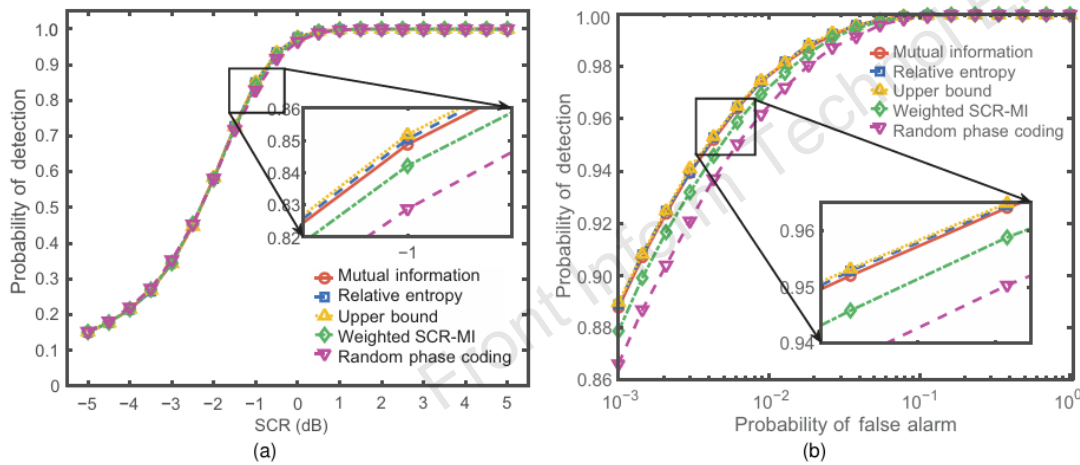


Fig. 2 Detection performance of waveforms using different criteria: (a) detection probability P_D vs. SCR ($P_{FA} = 10^{-2}$); (b) ROC curves (P_D vs. P_{FA}) at SCR = 0 dB

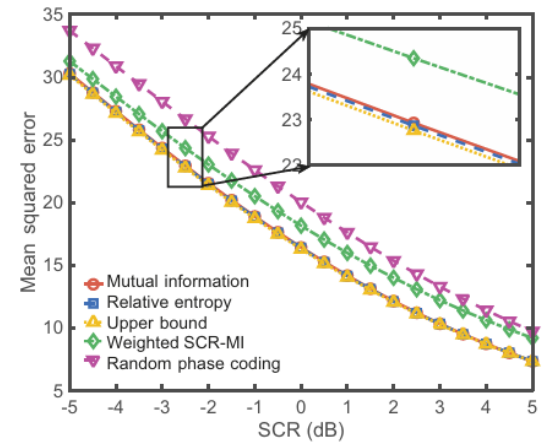


Fig. 3 MSE of TIR estimation vs. SCR for waveforms using different criteria

Major results (Cont'd)

Comparison of optimization objectives

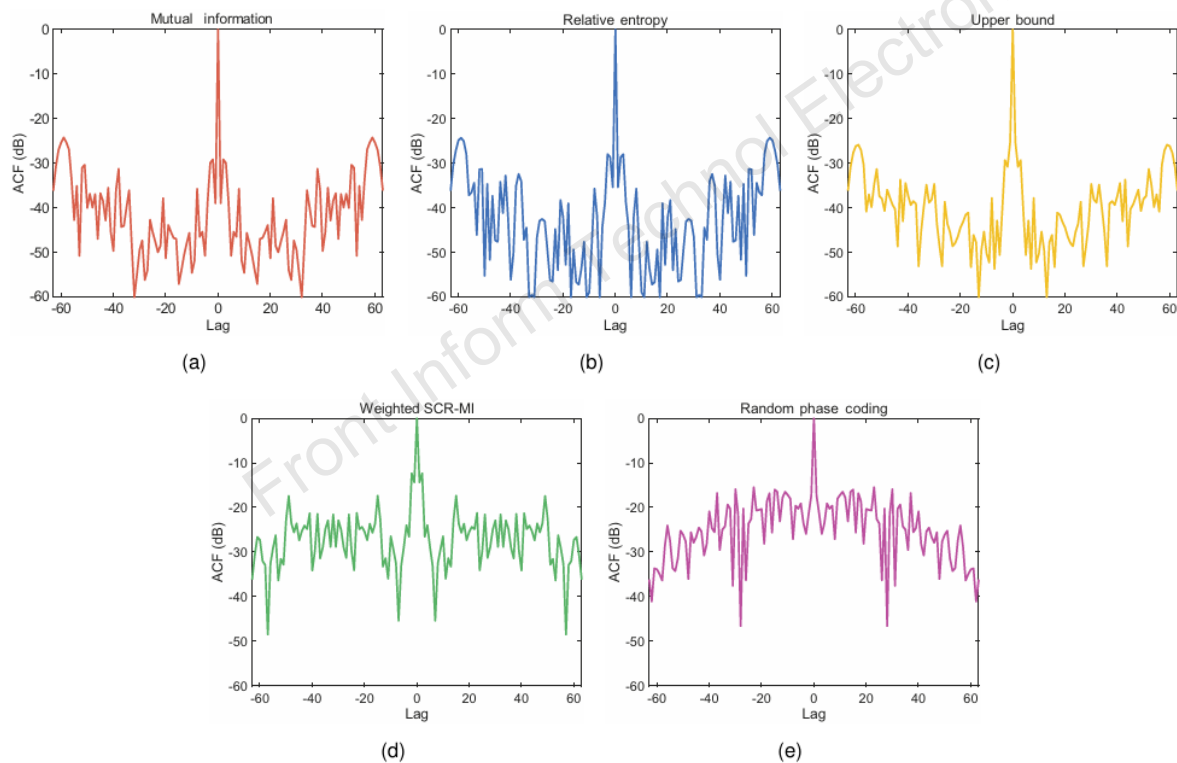


Fig. 4 Autocorrelation functions of waveforms using different criteria: (a) MI; (b) RE; (c) MIUB; (d) WSM; (e) RPC

Major results (Cont'd)

Comparison of optimization objectives

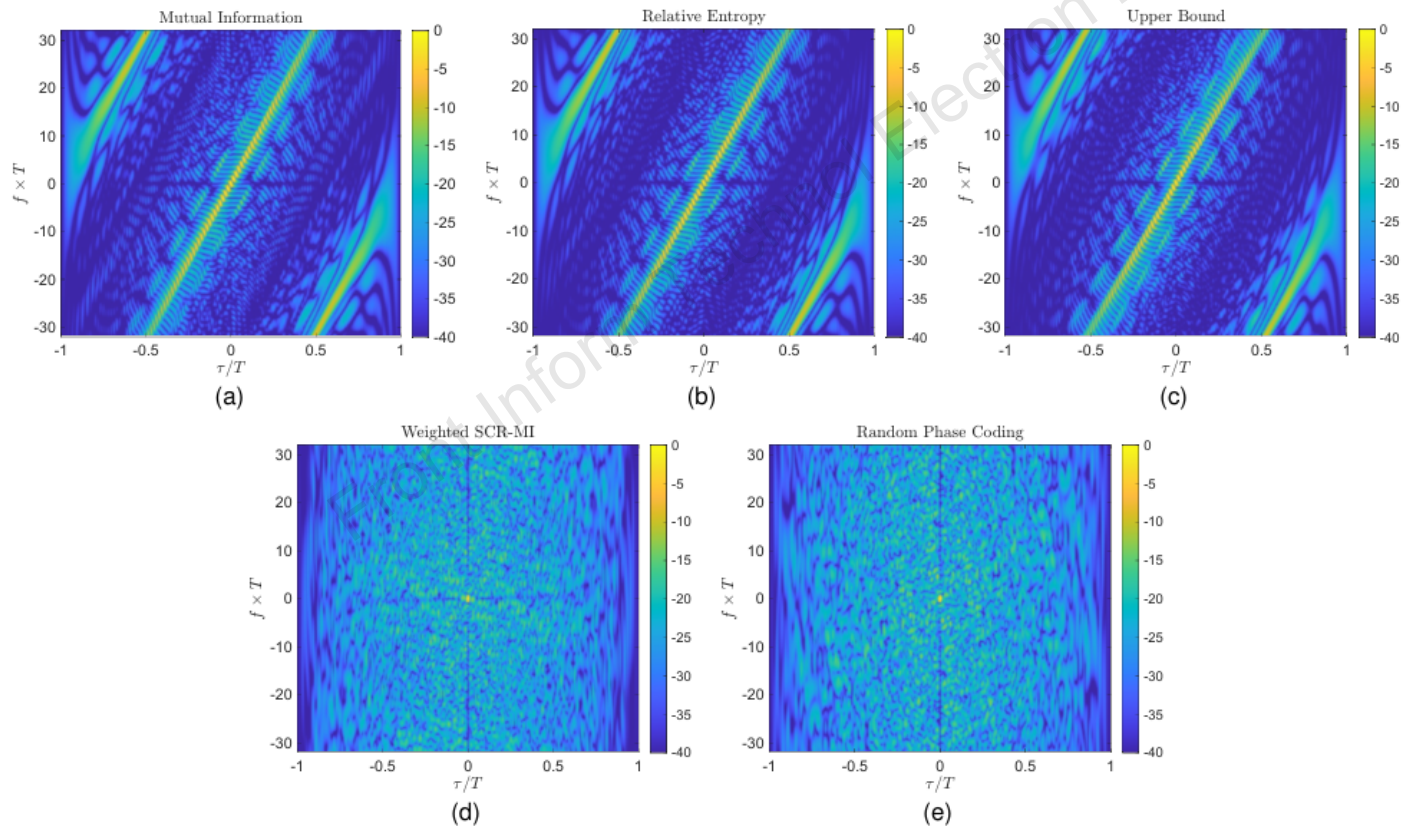


Fig. 5 Ambiguity functions of waveforms using different criteria: (a) MI; (b) RE; (c) MIUB; (d) WSM; (e) RPC

Major results (Cont'd)

Comparison of optimization algorithms

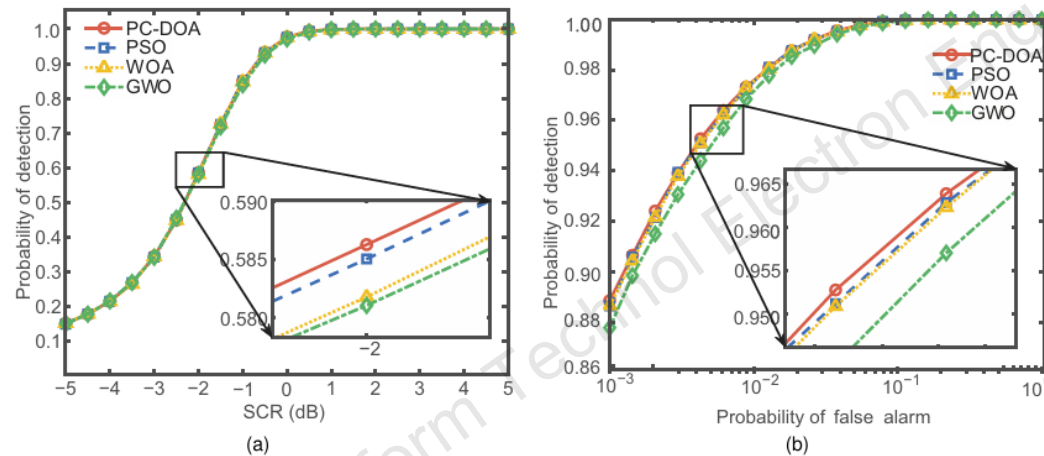


Fig. 6 Detection performance of waveforms using different optimization algorithms: (a) detection probability vs. SCR ($P_{FA} = 10^{-2}$); (b) ROC curves (P_D vs. P_{FA}) at SCR = 0 dB

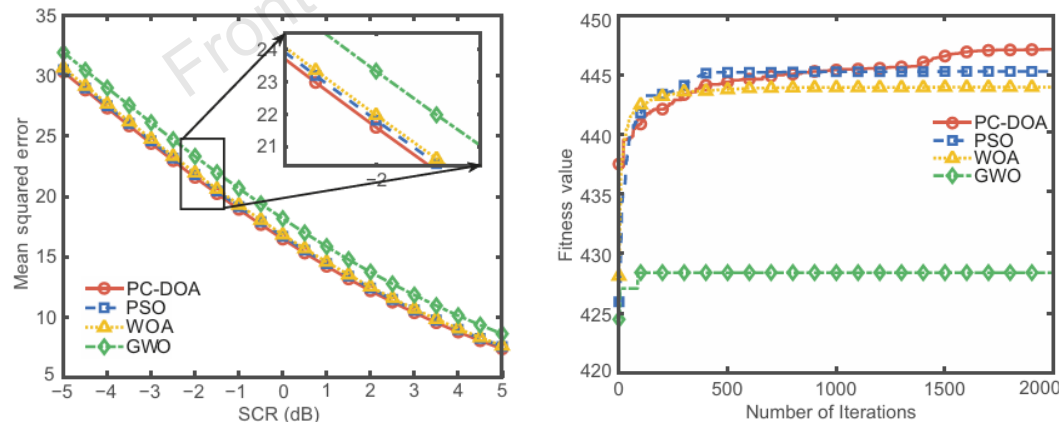


Fig. 7 MSE of TIR estimation vs. SCR of waveform using different optimization algorithms

Fig. 8 Convergence of best fitness $F(s)$ optimized using different optimization algorithms

Conclusions

1. This work presents a unified information-theoretic framework for constant modulus radar waveform design, jointly optimizing detection and estimation via an MIUB under Gaussian mixture models.
2. To address the ensuing nonconvex optimization, we propose PC-DOA, which leverages hybrid initialization and adaptive bi-phase search on the complex circle manifold.
3. Numerical evaluations confirm that MIUB-based designs outperform conventional baselines, offering superior detection-estimation trade-offs and desirable ambiguity function properties, thereby establishing a principled foundation for future extensions.



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